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**Nguyen et al.**

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(54) **SYSTEM AND METHOD FOR TRI-LEVEL LOGIC DATA SHUFFLING FOR OVERSAMPLING DATA CONVERSION**

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(52) **U.S. Cl.** ..... **341/144; 341/145**

(58) **Field of Classification Search** ..... **341/144, 341/145**

See application file for complete search history.

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*Primary Examiner*—Rexford Barnie

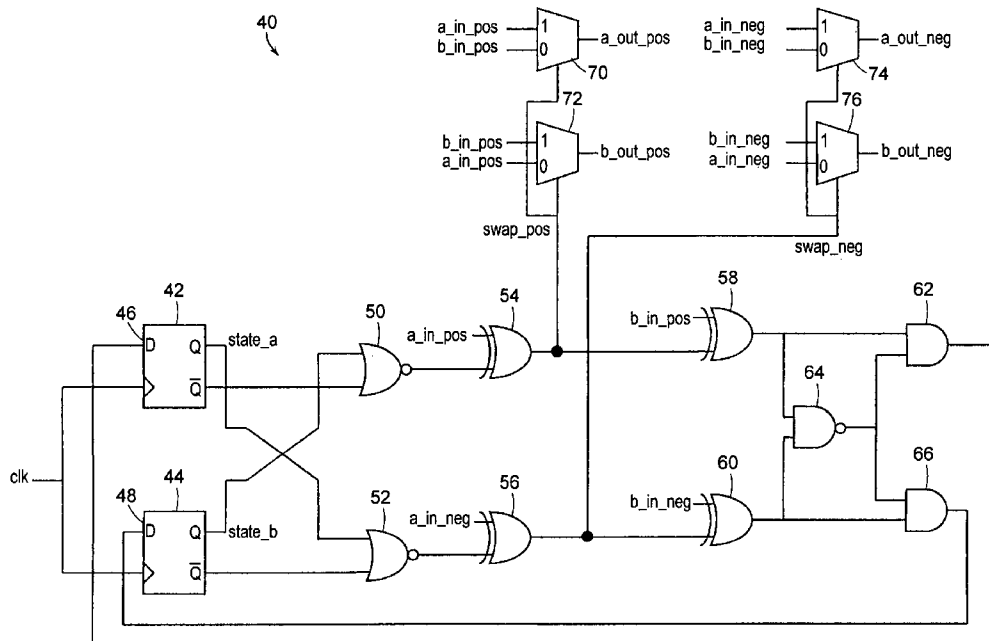
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(57) **ABSTRACT**

A system is disclosed for processing digital signals in a data converter. The system includes a thermometer encoder for receiving signed binary data and for providing signed thermometer data. The signed thermometer data includes positive thermometer data and negative thermometer data. The system also includes a shuffler that receives positive input data responsive to the positive thermometer data and receives negative input data responsive to the negative thermometer data. The system also includes a decoder for receiving output data from the shuffler and providing decoded data to an analog output stage.

**14 Claims, 12 Drawing Sheets**



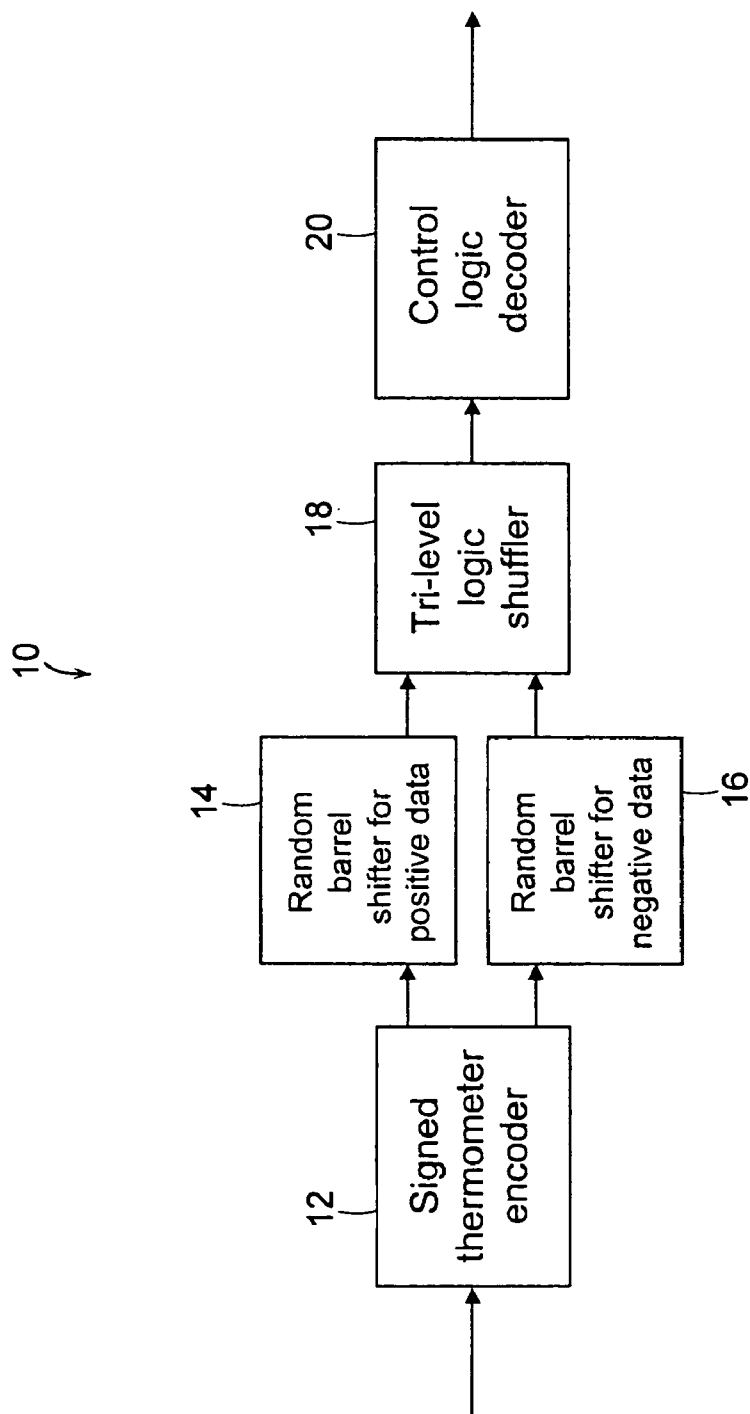


FIG. 1

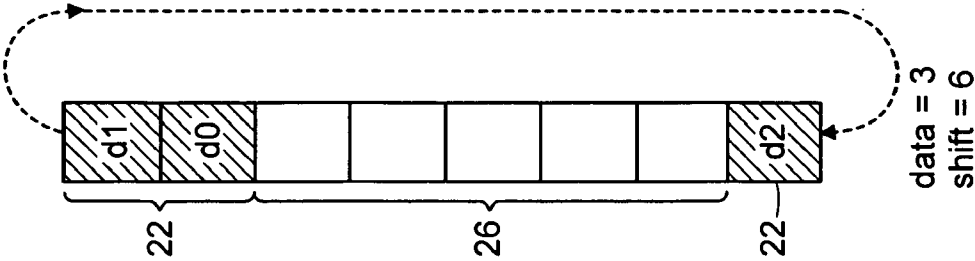


FIG. 2C

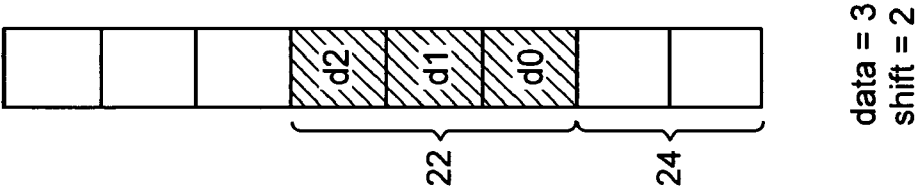


FIG. 2B

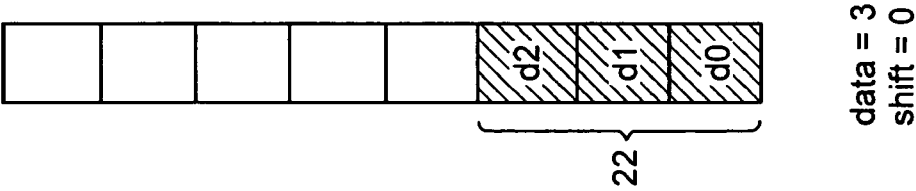


FIG. 2A

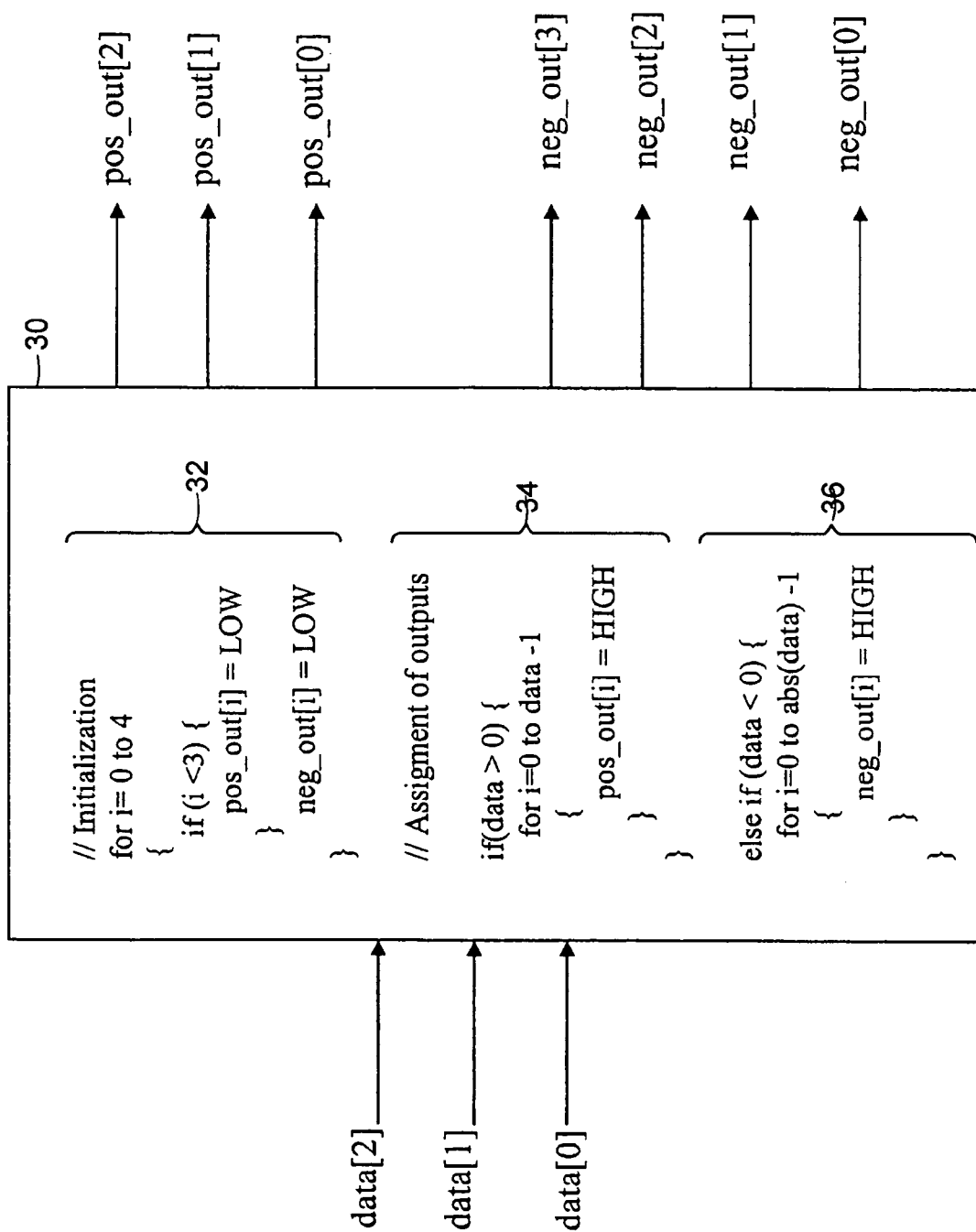
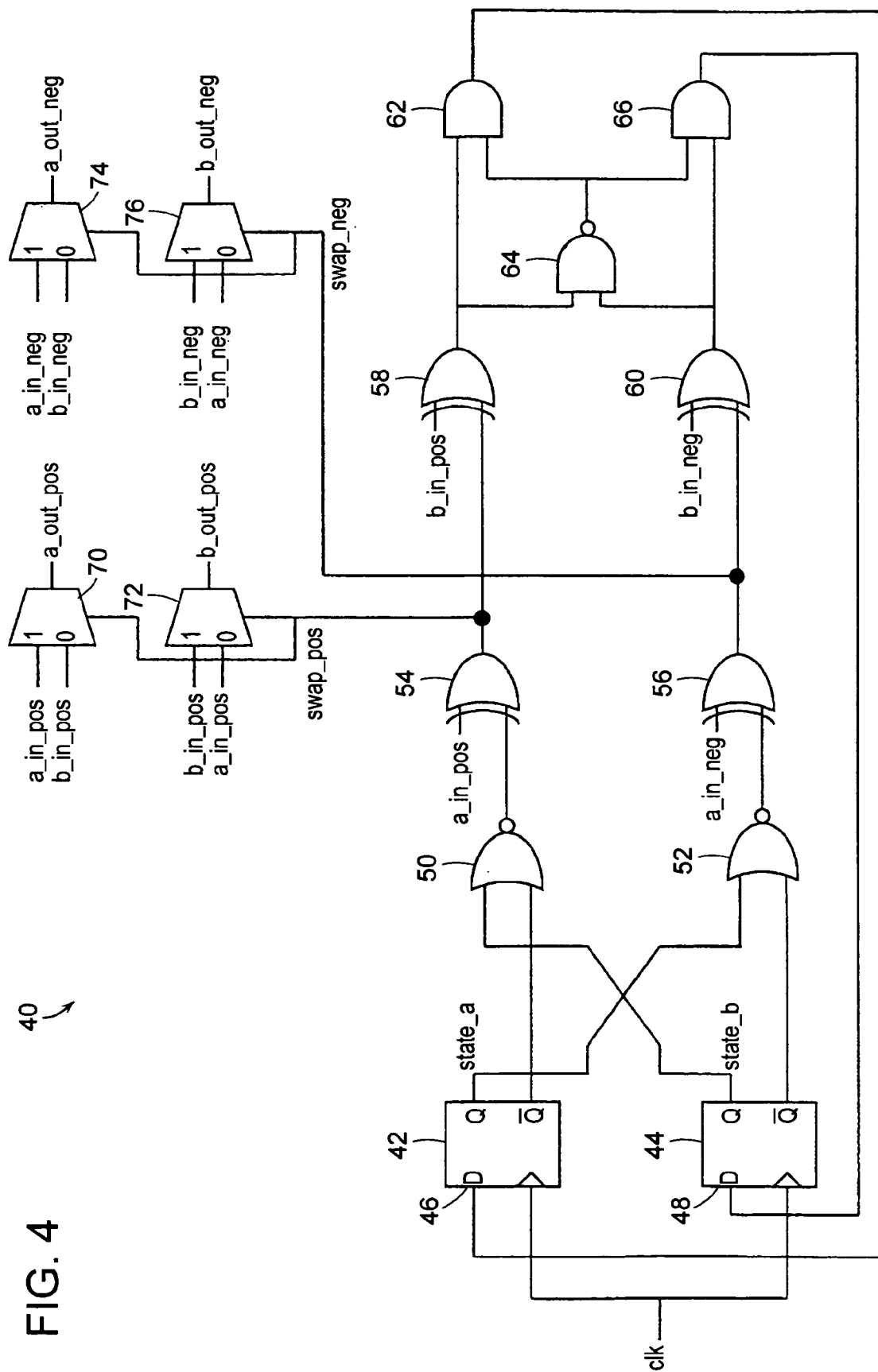


FIG. 3

40

FIG. 4



82 INPUTS					80 OUTPUTS					84	
<i>a_in_pos</i>	<i>b_in_pos</i>	<i>a_in_neg</i>	<i>b_in_neg</i>	current state <i>a</i>	current state <i>b</i>	<i>swap_pos</i>	<i>swap_neg</i>	next state <i>a</i>	next state <i>b</i>		
0	0	0	0	0	0	0	0	0	0		
0	0	0	0	0	1	0	0	0	1		
0	0	0	0	1	0	0	0	1	0		
0	0	0	1	0	0	0	0	0	1		
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1	1	1	1	1	1	0	0	0	1		

FIG. 5

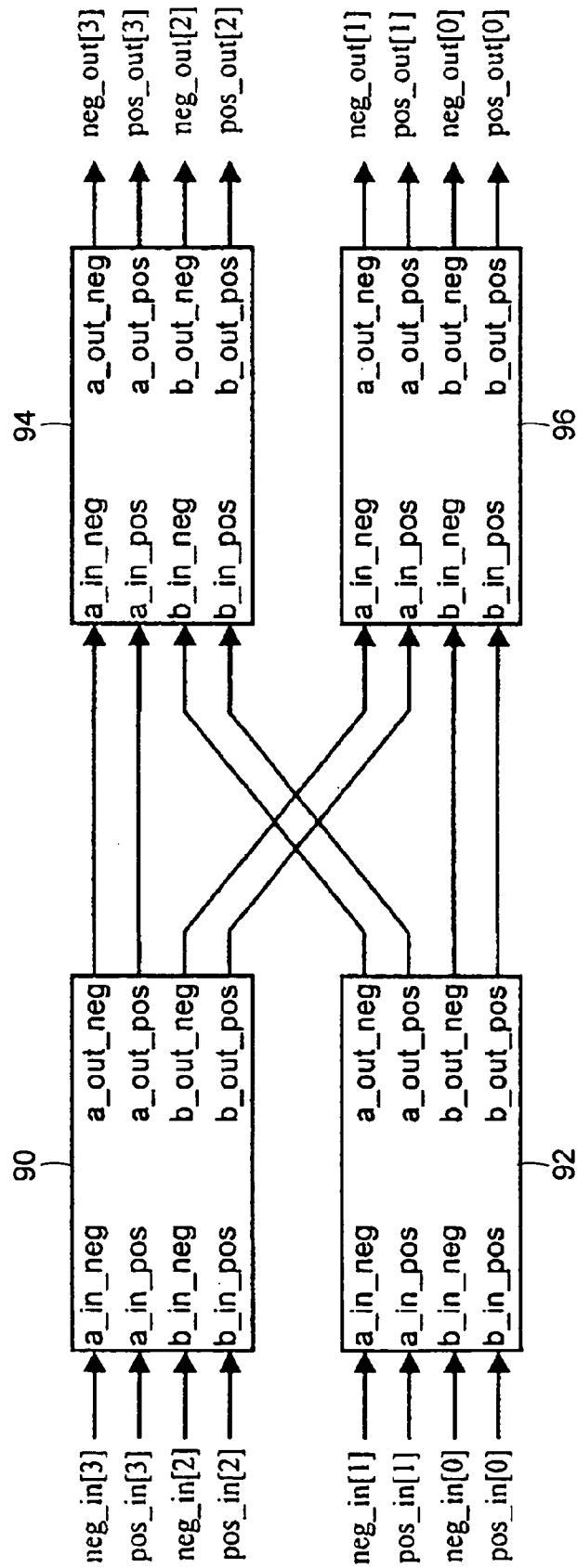


FIG. 6

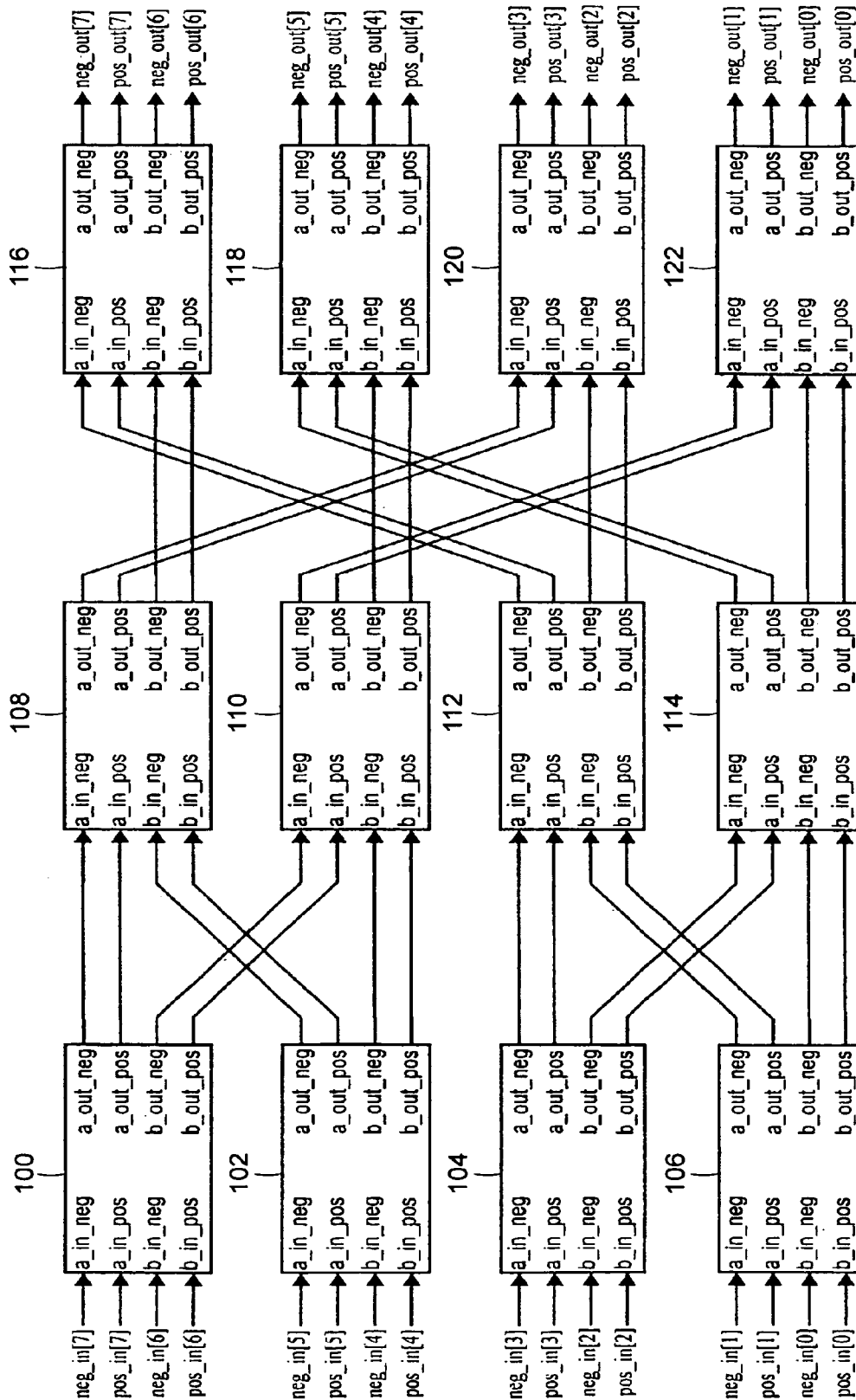
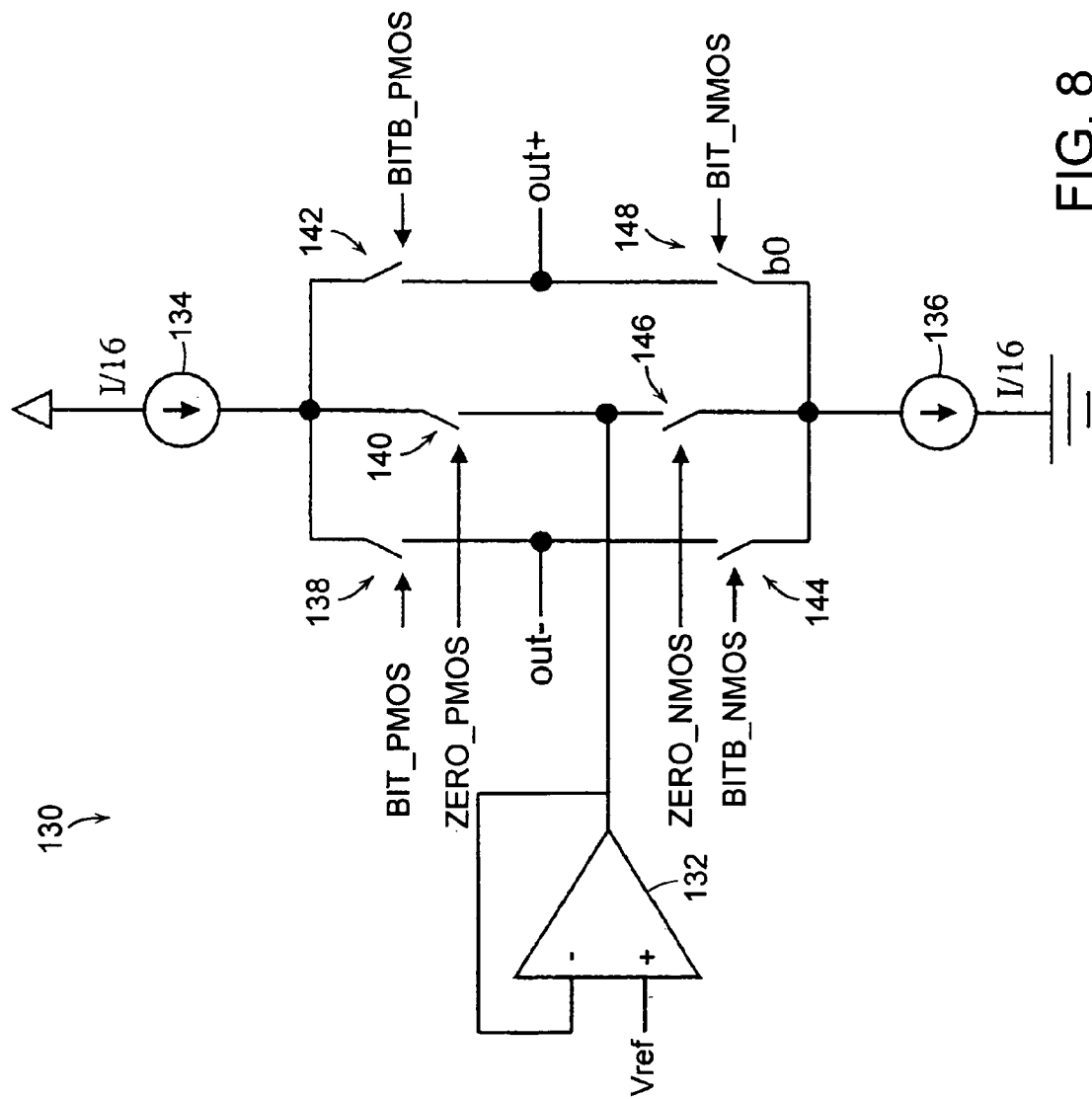


FIG. 7




$$\frac{\infty}{F/G}$$

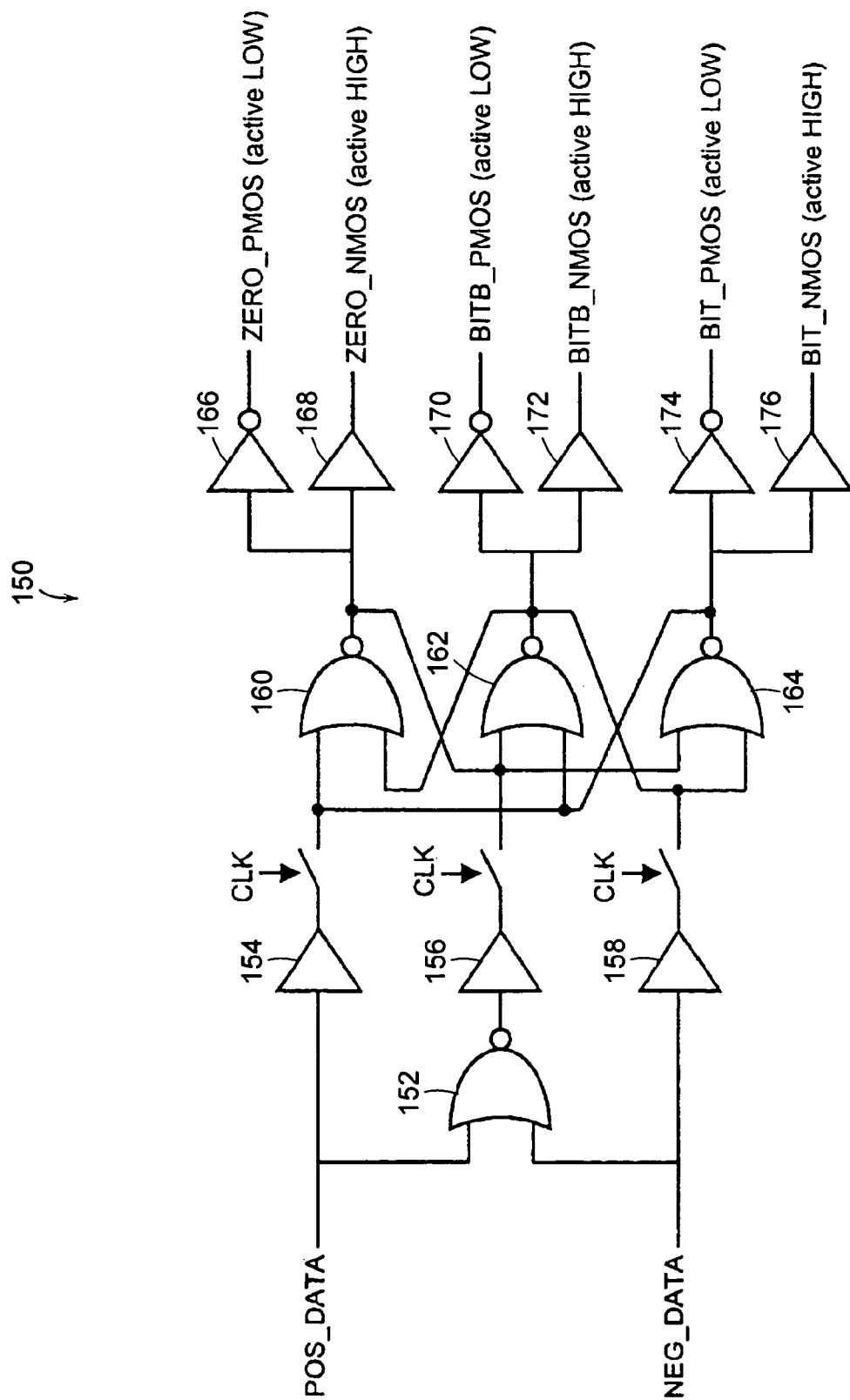


FIG. 9

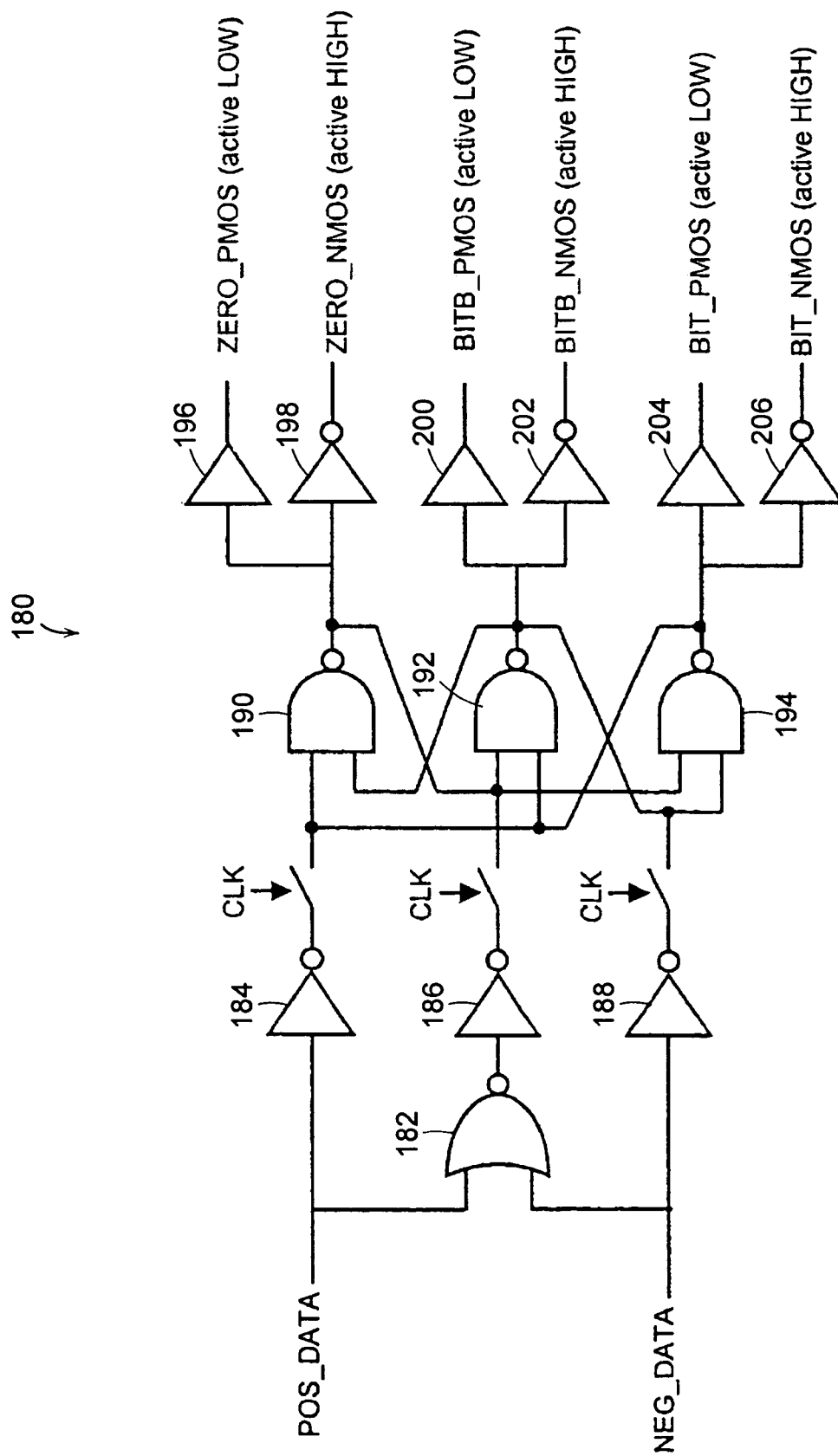


FIG. 10

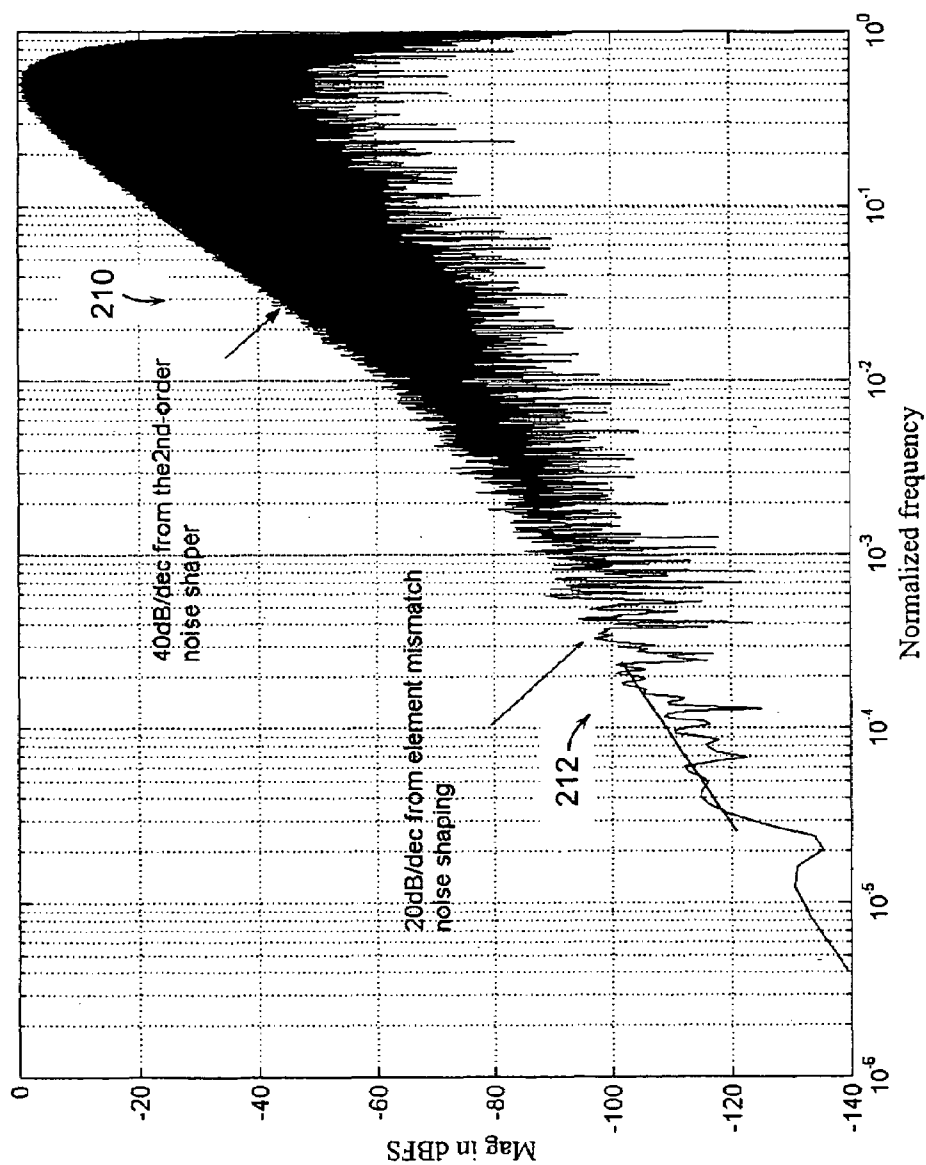


FIG. 11

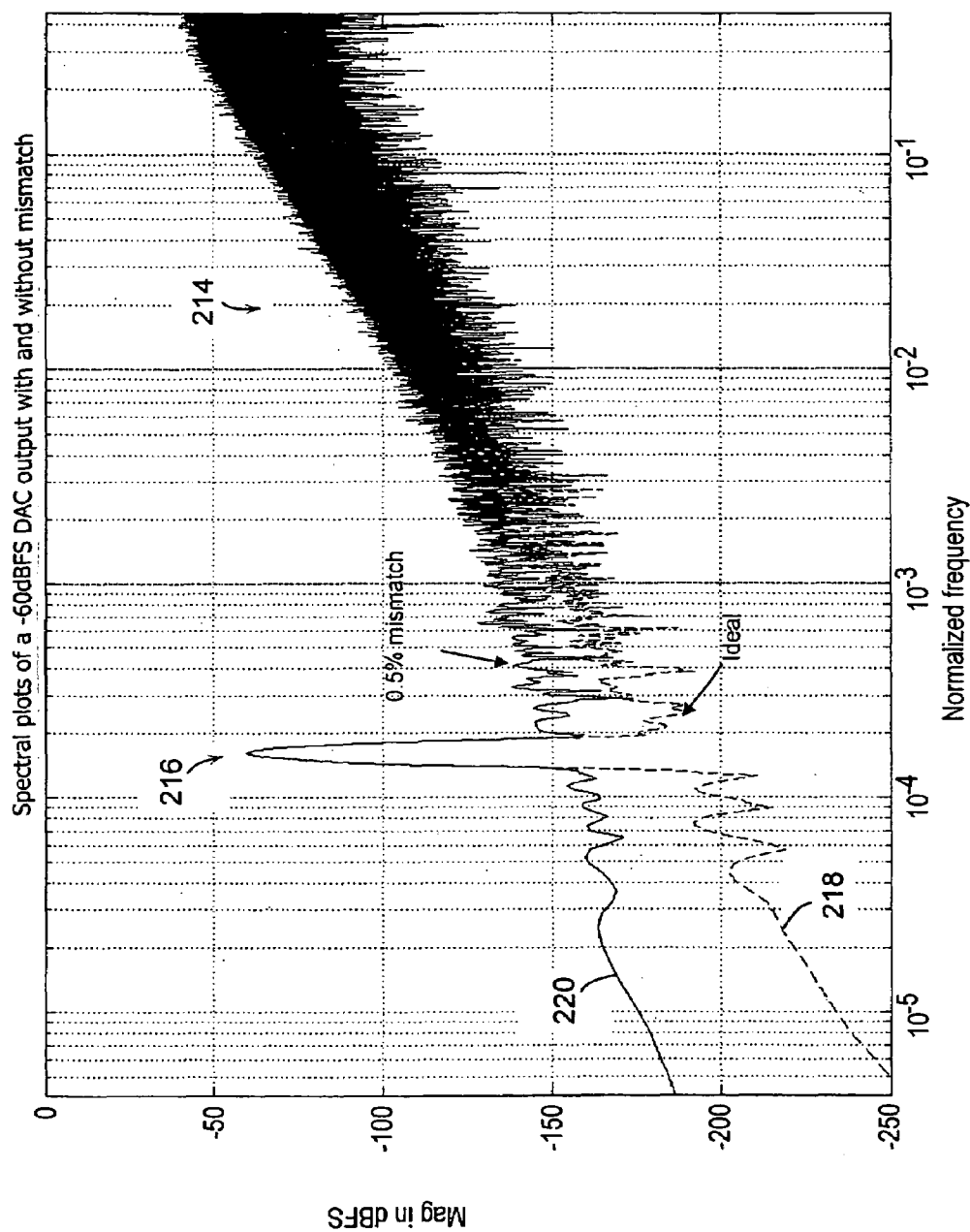


FIG. 12

1

# SYSTEM AND METHOD FOR TRI-LEVEL LOGIC DATA SHUFFLING FOR OVERSAMPLING DATA CONVERSION

## BACKGROUND OF THE INVENTION

The invention generally relates to mixed-signal converters of the sigma-delta noise shaping type, and more particularly, to mixed-signal digital-to-analog converters that employ uniformly weighted elements.

Sigma-delta digital-to-analog converters (DACs) provide for a means to achieve high resolution and low distortion at a relatively low cost compared to traditional Nyquist converters. In a typical multi-bit noise-shaped oversampling DAC, the digital input is first up-sampled by the oversampling ratio (OSR) and filtered to suppress the out-of-band images. A sigma-delta modulator is then used to reduce the word-width to a manageable size, and at the same time, shape the in-band noise to a higher frequency region. A binary to thermometer encoder is used to convert the binary data into thermometer-code data. For example, U.S. Pat. No. 5,404,142 discloses a data-directed scrambling technique in which a quantized noise-shaped word is first converted to a thermometer code. A data-directed shuffler is then used to dynamically select a group of elements of the output stage. The number of elements selected is equal to the number of active thermometer codes. An analog output stage then converts the output of the shuffler into an analog quantity by turning on the selected group of elements according to the decision of the shuffler.

A prior art thermometer-code DAC includes a current steering section and an I-to-V converter that includes a DAC cell driver that controls the BIT and BIT<sup>+</sup> (or BITB) signals. By designing the cross point of the BIT and BITB signals to be one  $V_{gs}$  above the common-mode voltage, the intersymbol interference (ISI) in the DAC cell output waveform will be minimized.  $V_{gs}$  is defined to be the gate-to-source voltage of the DAC switches when each is conducting half of the output current.

Due to imperfection of real devices, the current cells will not match exactly. This mismatch problem results in harmonic distortion and noise in the reconstructed analog signal. The performance of the converter is thus limited by the matching of these elements. Commercially available silicon processes can only offer matching of up to 12 bits without calibration or trimming.

This element mismatch has been well studied and methods have been proposed to convert the mismatch error into spectrally shaped noise. By shaping the mismatch error into out-of-band frequency region, the signal-to-noise ratio (SNR) and dynamic range (DNR) of the converter is greatly improved. In these methods, a shuffler (also sometimes called scrambler) is used to dynamically select a group of elements for every digital input code such that over time, each element is equally used. This implies that the first integral of the difference between every pair of elements is zero, hence, equivalent to a first-order noise shaped sigma-delta converters. The only difference is in a normal sigma-delta converter, the amplitude error is noise shaped whereas in a data shuffler, the error in the usage of the element is noise shaped. An example of a prior art butterfly-style shuffler is disclosed in U.S. Pat. No. 6,614,377. A drawback, however, of conventional thermometer-code current steering DACs is thermal noise performance. In particular, when the data is zero, half the number of the switching current sources are connected to one summing junction, and the other half are connected to the other summing junction of the I-to-V

2

converter. Moreover, the top current sources are always connected to the summing junctions. The current sources are the dominant noise source in the DAC output and dictate the SNR of the converter.

Another conventional DAC architecture that does not suffer from the mentioned noise problem includes tri-level logic thermometer current steering DAC that includes a pair of current sources (positive and negative) for each of the bits 0 to 15. Since each pair of current sources can be connected to the summing junction in three different ways, each pair may contribute a positive, a negative quantity of charge or nothing at all. When the data is zero, all the current sources are connected to a buffer amplifier to maintain their proper drain voltage. Therefore, the main noise source of the converter is now from the I-to-V amplifier, which by design, is much smaller than that of the current sources. Hence, the SNR is significantly improved. A difficulty with this architecture, however, is that it again may result in the element mismatch discussed above. Prior art element shufflers do not work with this architecture since they can only shuffle "1" and "0".

There is a need, therefore, for an improved sigma-delta noise-shaping DAC that further reduces element mismatch.

## SUMMARY

In accordance with an embodiment, the invention provides a system is disclosed for processing digital signals in a data converter. The system includes a thermometer encoder for receiving signed binary data and for providing signed thermometer data. The signed thermometer data includes positive thermometer data and negative thermometer data. The system also includes a shuffler that receives positive input data responsive to the positive thermometer data and receives negative input data responsive to the negative thermometer data. The system also includes a decoder for receiving output data from the shuffler and providing decoded data to an analog output stage.

In accordance with another embodiment, the invention provides a system for processing digital signals in a data converter that includes a thermometer encoder, a shuffler, and a decoder. The thermometer encoder is for receiving signed binary data and for providing signed thermometer data. The shuffler receives the signed thermometer data and provides tri-level logic output including positive state output, negative state output and zero state output. The decoder is for receiving output data from the shuffler and providing decoded data to an analog output stage.

In accordance with a further embodiment, the invention provides a method of processing digital signals in a data converter that includes the steps of receiving signed binary data at a thermometer encoder; providing signed thermometer data, receiving the signed thermometer data at a shuffler, providing tri-level logic output including positive state output, negative state output and zero state output; receiving output data from the shuffler at a decoder, and providing decoded data to an analog output stage.

## BRIEF DESCRIPTION OF THE DRAWINGS

The following description may be further understood with reference to the accompanying drawings in which:

FIG. 1 shows a diagrammatic illustrative view of a functional block diagram of a system in accordance with an embodiment of the invention;

FIGS. 2A–2C show diagrammatic illustrative functional depictions of 3-bit barrel shifters in accordance with an embodiment of the invention;

FIG. 3 shows a diagrammatic illustrative functional view of a 3-bit signed thermometer logic encoder in accordance with an embodiment of the invention;

FIG. 4 shows a diagrammatic illustrative view of a tri-level logic shuffling cell in accordance with an embodiment of the invention;

FIG. 5 shows an illustrative truth table for a shuffling cell in accordance with an embodiment of the invention;

FIG. 6 shows an illustrative diagrammatic view of a 3-bit tri-level logic shuffler in accordance with an embodiment of the invention;

FIG. 7 shows an illustrative diagrammatic view of 4-bit tri-level logic shuffler in accordance with another embodiment of the invention;

FIG. 8 shows an illustrative diagrammatic view of a DAC cell in accordance with an embodiment of the invention;

FIG. 9 shows an illustrative diagrammatic view of a NOR gate-based DAC cell driver in accordance with an embodiment of the invention;

FIG. 10 shows an illustrative diagrammatic view of a NAND gate-based DAC cell driver in accordance with another embodiment of the invention;

FIG. 11 shows an illustrative diagrammatic graphical representation of a simulation for a 4-bit  $2^{nd}$  order DAC with 0.5% element mismatch; and

FIG. 12 shows an illustrative diagrammatic graphical representation of the simulation of FIG. 11 with a -60 dBFS input.

The drawings are shown for illustrative purposes only.

#### DETAILED DESCRIPTION OF THE ILLUSTRATED EMBODIMENTS

The invention provides a shuffler that can handle “+1”, “0” and “-1”, or tri-level logic data, and in particular provides a DAC cell driver that can generate three signals driving the BIT, BITB and the ZERO in such a way that the ISI in the output waveform is minimized.

The present invention provides for a tri-level logic thermometer (or signed thermometer) encoder, a shuffler, a control logic decoder, and a DAC cell driver. The function of the tri-level logic thermometer encoder is to encode signed binary data into signed thermometer code. The function of the control logic decoder is to decode the output of the shuffler into control signals for the output stage. The function of the shuffler is to shape the mismatch error into an out-of-band frequency region. FIG. 1 shows the block diagram of a system 10 in accordance with an embodiment of the present invention.

As shown in FIG. 1, signed binary data is received by a signed binary thermometer encoder 12 that provides signed thermometer data to both a random barrel shifter for positive data 14, and a random barrel shifter for negative data 16. The output of each of the random barrel shifters 14 and 16 is provided to a tri-level logic shuffler 18, and the signed shuffled thermometer data that is output by the shuffler 18 is provided to a control logic decoder 20, which provides control signals to the DAC.

The mismatch error of an element  $u_i$  is defined as the difference between the actual value of element and the average value of all the elements. Specifically, for an N-element thermometer DAC, the error of element  $u_i$  is:

$$e_i = u_i - \frac{1}{N} \sum_{k=0}^{N-1} u_k$$

Each time when the data is “+1” and the element  $u_i$  is selected, the error contributed to the output is  $+e_i$ . Each time when the data is “-1” and the element  $u_i$  is selected, the error contributed to the output is  $-e_i$ . When the data is zero, the error contributed by element  $u_i$  during that particular clock cycle is zero.

The shuffler has two functions. First it selects the elements so that the average contribution of the error of each element is zero. Second, the average usage of each element compared to all other elements is kept equal. The first function is achieved by the a shuffler cell in accordance with an embodiment of the present invention while the second function is achieved by the butterfly connection in the overall shuffler. By shuffling the usage of the elements in a data-directed manner, the mismatch error is converted to noise and shaped to out-of-band frequency region.

The present invention also provides for two barrel shifters that are controlled by a random number generator. These barrel shifters act to de-correlate the digital input signal of the shuffler so that idle tones are not generated by the shuffler cells. FIGS. 2A, 2B and 2C show the operation of a 3-bit random barrel shifter when the data is 3 and the shift is 0, 2 and 6 respectively. For example, in FIG. 2A the data is a code of 3 as shown at 22, and the shift control, which is generated by a pseudo-random generator, is 0, the data will occupy the bottom 3 bits. When the shift control is greater than 0, data will be shifted up by the corresponding number of positions (e.g., 2 or 6 as shown in at 24 and 26 in FIGS. 2B and 2C) and will wrap around if necessary as shown in FIG. 2C.

FIG. 3 shows the 3-bit signed thermometer logic encoder 30 with a descriptive depiction of the code provided therein for illustrative purposes. The input data is signed binary data (including a sign bit data[2] and a magnitude bit data[1] as well as a zero bit data[0]), while the output data is signed thermometer data (including pos\_out[0], pos\_out[1], pos\_out[2], neg\_out[0], neg\_out[1], neg\_out[2] and neg\_out[3]). As shown in the code at 32, both the positive outputs and negative outputs are first initialized to 0, then depending on the input data value, either the positive output bits or negative output bits are asserted 1 as shown at 34 and 36 respectively. There is never the case where both positive and negative outputs are asserted since this represents an illegal condition.

FIG. 4 shows a schematic diagram of a shuffler cell 40 in accordance with an embodiment of the invention. The shuffler cell 40 includes D flip flops 42 and 44 that receive a clock input signal and current\_state\_a and current\_state\_b signals at inputs 46 and 48 respectively. The Q outputs of each of the flip flops 42 and 44 are provided to NOR gates 50 and 52 together with the  $\bar{Q}$  outputs of the other respective flip flop as shown. The Q output of the flip flop 42 provides the next\_state\_a and the Q output of the flip flop 44 provides the next\_state\_b. The output of the NOR gate 50 is provided to an exclusive OR (XOR) gate 54 together with the a\_in\_pos\_data, and the output of the NOR gate 52 is provided to an XOR gate 56 together with the a\_in\_neg\_data. The output of the XOR gate 54 is provided to an XOR gate 58 together with the b\_in\_pos\_data, and the output of the XOR gate 56 is provided to an XOR gate 60 together with the b\_in\_neg\_data.

5

data. The output of the XOR gate **58** is provided to both an AND gate **62** and to a NAND gate **64**. The output of the XOR gate **60** is provided to an AND gate **66** and is also provided to the NAND gate **64**. The output of the NAND gate **64** is provided to inputs of both of the AND gates **62** and **66**. The outputs of the AND gates **62** and **66** are provided to the D inputs **46** and **48** of the flip flops **42** and **44** respectively.

The a/b equalization for the positive and negative circuits is provided by summers **72**, **74**, **76** and **78**. In particular, the output of the XOR gate **54** provides a swap\_pos signal to clock signals of the summers **70** and **72**, and the XOR gate **56** provides a swap\_neg signal to the clock signals of the summers **74** and **76**.

The operation of the shuffler cell **40** is as follows. On the rising edge of the clock the state variable state\_a and state\_b are updated. Variable state\_a records which of the 2 positive outputs was used previously while state\_b records which of the 2 negative outputs was used previously. Specifically, if state\_a is 1, a\_out\_pos was used previously. Likewise, if state\_b is 1, a\_out\_neg was used previously. When both state\_a and state\_b are 1s, they indicate that a\_out\_pos and a\_out\_neg were both used previously, this means that the error contributed by the element connected to a\_out\_pos and a\_out\_neg is averaged out and equal to zero.

The truth table for valid values of the logic circuit for the shuffler cell **40** of FIG. **4** is shown at **80** in FIG. **5**. Although there are 64 total combinations of inputs, certain combinations are not permitted since they have both positive and negative inputs being 1s' at the same time, or state\_a and state\_b are both equal to 1s' at the same time. The valid input possible combinations are shown at **82** and the valid possible output combinations are shown at **84**.

FIG. **6** shows a 3-bit tri-level logic shuffler that receives eight lines of input and produces eight lines of output. The cell employs a butterfly connection in which pairs of data lines are cross coupled as shown. In particular, circuit includes a swap cell **90** that receives input data neg\_in[3], pos\_in[3], neg\_in[2] and pos\_in[2]. The swap cell **92** receives input data neg\_in[1], pos\_in[1], neg\_in[0] and pos\_in[0]. The a\_out\_neg data and the a\_out\_pos data from the cell **90** are transferred directly a\_in\_neg and a\_in\_pos inputs of a cell **94**, while the b\_out\_neg data and the b\_out\_pos data from the cell **90** are crossed to a\_in\_neg and a\_in\_pos inputs of a cell **96**. The a\_out\_neg data and the a\_out\_pos data from the cell **90** are crossed to the b\_in\_neg and b\_in\_pos inputs of a cell **94**, while the b\_out\_neg data and the b\_out\_pos data from the cell **90** are transferred directly to b\_in\_neg and b\_in\_pos inputs of cell **96**.

FIG. **7** shows a 4-bit tri-level logic shuffler circuit that receives eight pairs of inputs and produces 8 pairs of outputs. The circuit includes swap cells **100**, **102**, **108** and **110** that function as discussed above with respect to the circuit of FIG. **6**. The circuit also includes cells **104**, **106**, **112** and **114** that also function as discussed above with respect to FIG. **6**. A second stage is also provided in the circuit of FIG. **7** in which the b\_out\_neg and b\_out\_pos data are provided directly to the cells **116**, **118**, **120** and **122** respectively, but the a\_out\_neg and a\_out\_pos data are crossed. In particular, the a\_out\_neg and a\_out\_pos data of the cell **108** are crossed to the a\_in\_neg and a\_in\_pos data inputs of the cell **120**; the a\_out\_neg and a\_out\_pos data of the cell **110** are crossed to the a\_in\_neg and a\_in\_pos data inputs of the cell **122**; the a\_out\_neg and a\_out\_pos data of the cell **112** are crossed to the a\_in\_neg and a\_in\_pos data inputs of the cell **116**; and the a\_out\_neg and a\_out\_pos data of the cell **114** are crossed to the a\_in\_neg and a\_in\_pos data

6

inputs of the cell **118**. Since each shuffler has 4 outputs, it will be connected, therefore, to 2 pairs of push-pull current cells.

FIG. **8** shows a DAC cell **130** in accordance with an embodiment of the invention that includes an amplifier **132**, a positive current source **134**, a negative current source **136**, and switches **138**, **140**, **142**, **144**, **146** and **148** that are employed to provide the BIT\_PMOS, BITB\_PMOS, BITB\_NMOS and BIT\_MOS outputs as shown.

FIG. **9** shows a DAC cell **150** in accordance with an embodiment of the invention that employs a NOR gate based design. In particular, the circuit **150** includes NOR gates **152**, **160**, **162** and **164**, as well as buffers **154**, **156**, **158**, **168**, **172** and **176**, and inverters **166**, **170** and **174**. FIG. **10** shows a DAC cell **180** in accordance with another embodiment of the invention that employs a NAND gate based design. In particular the circuit **180** includes a NOR gate **182** and NAND gates **190**, **192** and **194**, as well as buffers **184**, **186**, **188**, **196**, **200**, **204**, as well as inverters **198**, **202** and **206**.

FIG. **11** shows spectral plot of the output of a 2<sup>nd</sup>-order, 4-bit DAC in accordance with a tri-level logic data-directed shuffler of an embodiment of the invention. The DAC has 16 unit elements that have a mismatch error of 0.5% rms. FIG. **11** shows the 40 dB/dec from the 2<sup>nd</sup> order noise shaper at **210** and the 20 dB/dec from element mismatch noise shaping at **212**. FIG. **12** shows a -60 dBFS output of the same converter. In particular, FIG. **12** shows the -100 to -50 dBFS at **214** with a peak shown at **216**, with the simulation results with element mismatch shown at **220** and the ideal results without element mismatch shown at **218**. Both FIGS. **11** and **12** show the noise shaping effect of the shuffler as a 20 dB/dec slope of the spectral plots.

In various embodiments, therefore, the invention provides a method of generating a signed thermometer data output from a signed binary data input in which the method includes the step of generating two thermometer data outputs; one responding to positive input data and one responding to negative input data. In further embodiments, the invention provides a method of pre-randomizing the positive and negative thermometer data groups, and may further include the steps of generating a pseudo-random number; and using the pseudo-random number to control two barrel shifters to shift the position of the thermometer data words within the barrel shifters.

In accordance with a further embodiment, the invention provides a method of grouping signed thermometer data before they enter a shuffler that includes the steps of paring the positive thermometer data into groups of two; and pairing the negative thermometer data into groups of two.

In accordance with another embodiment, the invention provides a method of shuffling signed thermometer data that includes the steps of generating a shuffler cell; generating the state variables, and a swap control signal within said shuffler cell based on the previous state variables and the current inputs; connecting the inputs of the shuffler cell directly to the outputs; or swapping the inputs before connecting them to the outputs according to the value of the swap control signal.

In accordance with another embodiment, the invention provides a signed thermometer data shuffler cell that includes a pair of positive thermometer data inputs, a pair of negative thermometer data inputs, a pair of positive thermometer data outputs, a pair of negative thermometer data outputs, and a clock signal and a reset signal.

In accordance with another embodiment, the invention further provides that the positive output pair of the previous



7

shuffler cell may be connected to the positive input pair of the following shuffler cell; and that the negative output pair of the previous shuffler cell may be connected to the negative input pair of the following shuffler cell; wherein both connections follow the butterfly-style connection discussed above.

In accordance with another embodiment, the invention provides a method of decoding the output of the shuffler that includes the steps of generating a new control signal from the positive and negative inputs where the control signal is asserted when both inputs are low; connecting the control signal to a pair of switches; connecting the positive input to a pair of switch; and connecting the negative input to a pair of switches.

In accordance with yet another embodiment, the invention provides a method of holding the drain voltage of the current sources at a known level that includes the step of connecting the current sources to a reference voltage buffer output when the input control signals are both zero.

Those skilled in the art will appreciate that numerous variations and modifications may be made to the above disclosed embodiments without departing from the spirit and scope of the invention.

What is claimed is:

1. A system for processing digital signals in a data converter, said system comprising:

a thermometer encoder for receiving signed binary data and for providing signed thermometer data, said signed thermometer data including positive thermometer data and negative thermometer data;

a shuffler that receives positive input data responsive to the positive thermometer data and that receives negative input data responsive to the negative thermometer data;

a decoder for receiving output data from the shuffler and providing decoded data to an analog output stage; and

a positive data barrel shifter interposed between the thermometer encoder and the shuffler, and a negative barrel shifter interposed between the thermometer encoder and the shuffler.

2. The system as claimed in claim 1, wherein said thermometer encoder includes a first positive output data node for providing positive thermometer data and a negative output node for providing negative.

3. The system as claimed in claim 1, wherein said shuffler includes a plurality of shuffler cells, each of which receives two pairs of inputs and provides two pairs of outputs, wherein one of said pairs of outputs is coupled to a second shuffler cell while a second of said pairs of outputs is coupled to a third shuffler cell.

4. The system as claimed in claim 3, wherein said shuffler receives four pairs of data inputs and provides four pairs of data outputs.

5. The system as claimed in claim 3, wherein each said shuffler cell includes a pair of positive thermometer data inputs, a pair of negative thermometer data inputs, a pair of

8

positive thermometer data outputs, a pair of negative thermometer data outputs, a clock signal and a reset signal.

6. The system as claimed in claim 1, wherein said analog output stage includes a tri-level logic output driver that provides a positive output, a negative output and a zero output.

7. The system as claimed in claim 6, wherein each said shuffler cell includes flip flops.

8. A system for processing digital signals in a data converter, said system comprising:

a thermometer encoder for receiving signed binary data and for providing signed thermometer data;

a shuffler that receives said signed thermometer data and provides tri-level logic output including positive state output, negative state output and zero state output;

a decoder for receiving output data from the shuffler and providing decoded data to an analog output stage; and a positive data barrel shifter interposed between the thermometer encoder and the shuffler, and a negative barrel shifter interposed between the thermometer encoder and the shifter.

9. The system as claimed in claim 8, wherein said signed thermometer data includes positive thermometer data and negative thermometer data.

10. The system as claimed in claim 8, wherein said shuffler includes a plurality of shuffler cells, each of which receives two pairs of inputs and provides two pairs of outputs, wherein one of said pairs of outputs is coupled to a second shuffler cell while a second of said pairs of outputs is coupled to a third shuffler cell.

11. The system as claimed in claim 10, wherein said shuffler receives four pairs of data inputs and provides four pairs of data outputs.

12. The system as claimed in claim 10, wherein each said shuffler cell includes a pair of positive thermometer data inputs, a pair of negative thermometer data inputs, a pair of positive thermometer data outputs, a pair of negative thermometer data outputs, a clock signal and a reset signal.

13. The system as claimed in claim 8, wherein said shuffler provides that current sources are coupled to a reference voltage when said zero state output is provided by the shuffler.

14. A method of processing digital signals in a data converter, said method comprising the steps of:

receiving signed binary data at a thermometer encoder; providing signed thermometer data;

receiving said signed thermometer data at a shuffler;

providing tri-level logic output including positive state output, negative state output and zero state output, when current sources are coupled to a reference voltage;

receiving output data from the shuffler at a decoder; and providing decoded data to an analog output stage.

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